

# CS4140

## Embedded Systems Laboratory

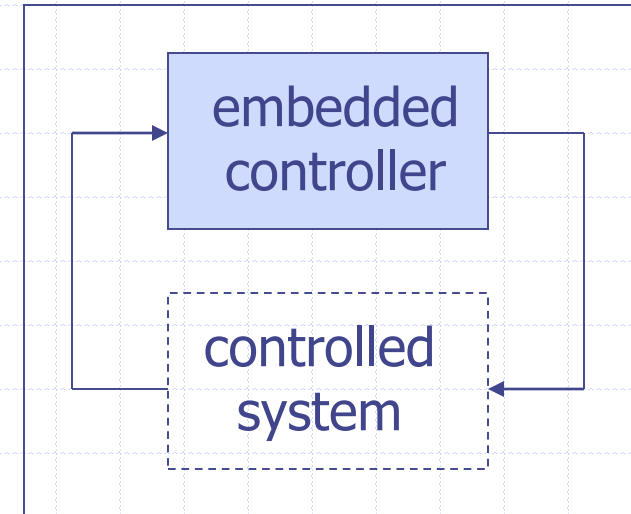
### Introduction to Control Theory

# Why Control Theory?

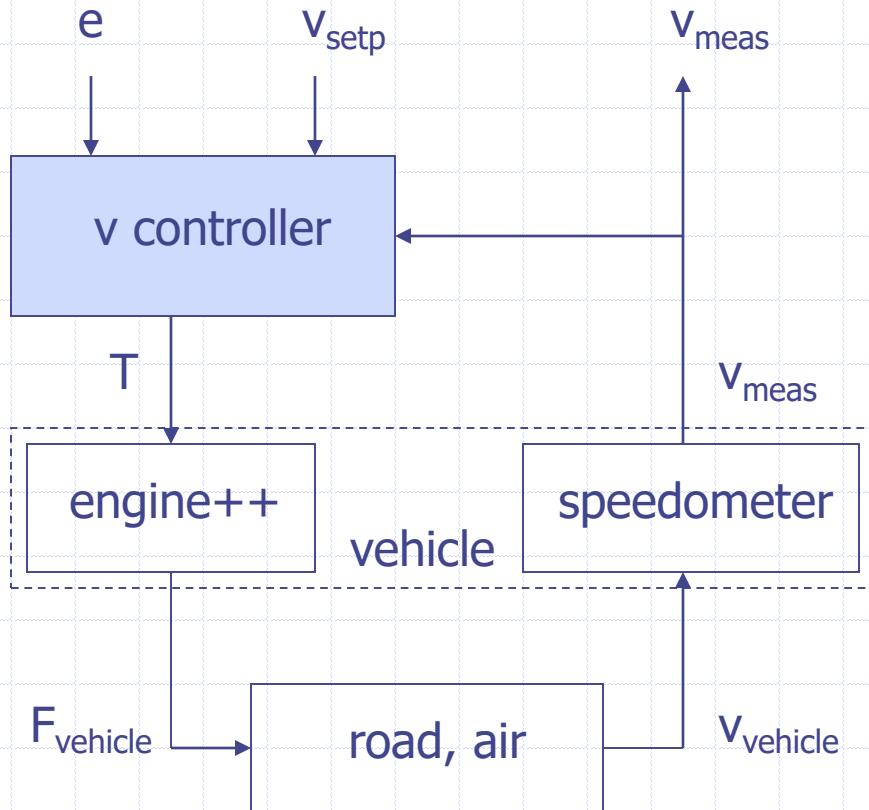
- ◆ Embedded systems integrated with appl'n
- ◆ Multi-disciplinary training required:
  - Physics engineering
  - Electronics engineering
  - Mechanical engineering
  - **Control engineering**
  - ...
  - And, of course,
  - Computer science & engineering

# Control is Everywhere

- ◆ Automotive
- ◆ Aerospace
- ◆ Plant Control
- ◆ Climate Control
- ◆ Health Care
- ◆ Copiers, Wafer Scanners
- ◆ Model Quad Rotors ...



# Cruise Control



$e$  = enable [0/1]  
 $T$  = throttle [%]  
 $F$  = thrust [N]  
 $v$  = velocity [m/s]

$V_{\text{setp}}$  = setpoint  
 $V_{\text{meas}}$  = measured  
 $V_{\text{vehicle}}$  = actual

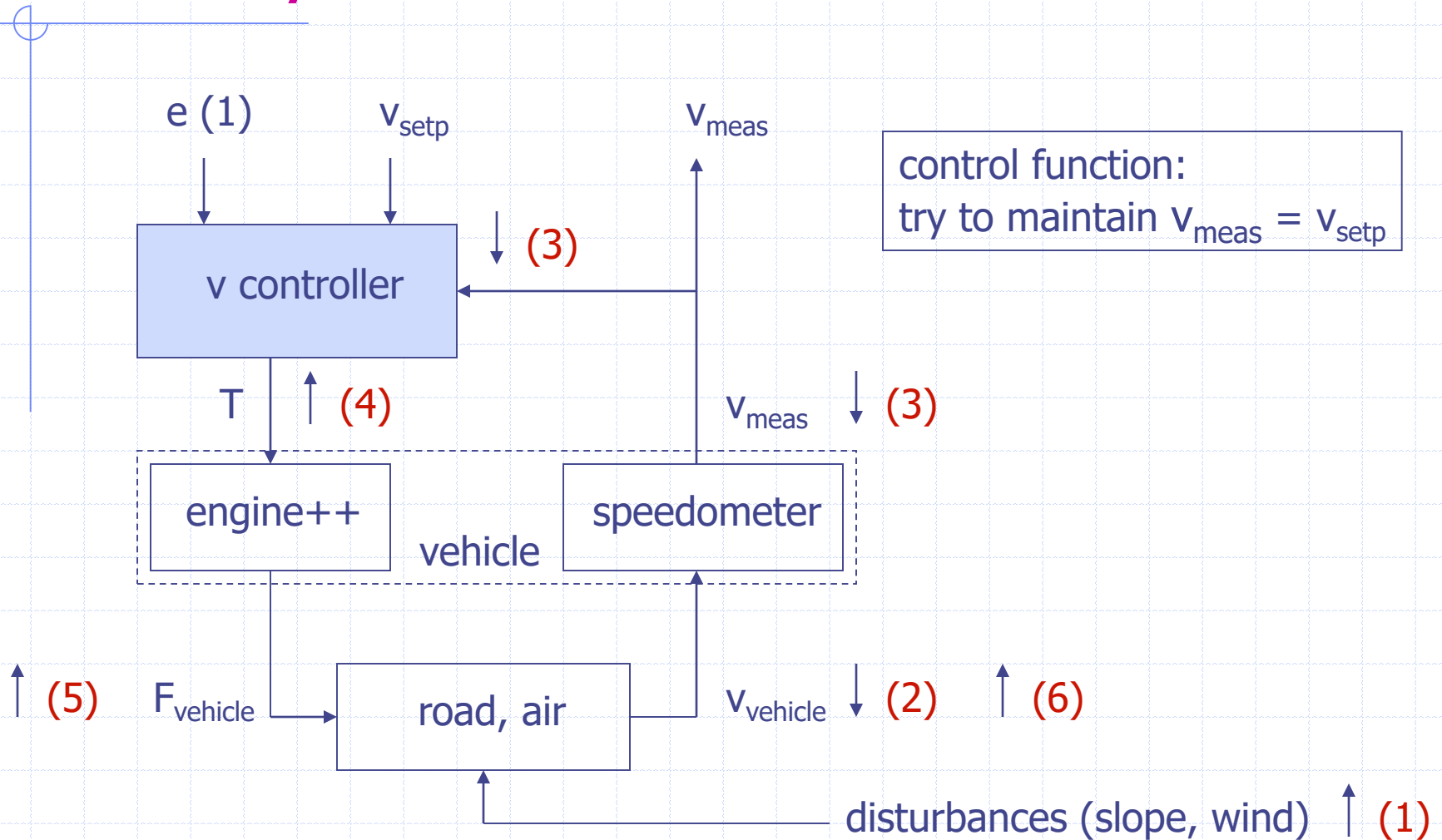
# Objectives of this Crash Course

- ◆ Appreciate the benefits of control
- ◆ Understand basic control principles
- ◆ Communicate with control engineers
  
- ◆ Get you up to speed to do the QR control

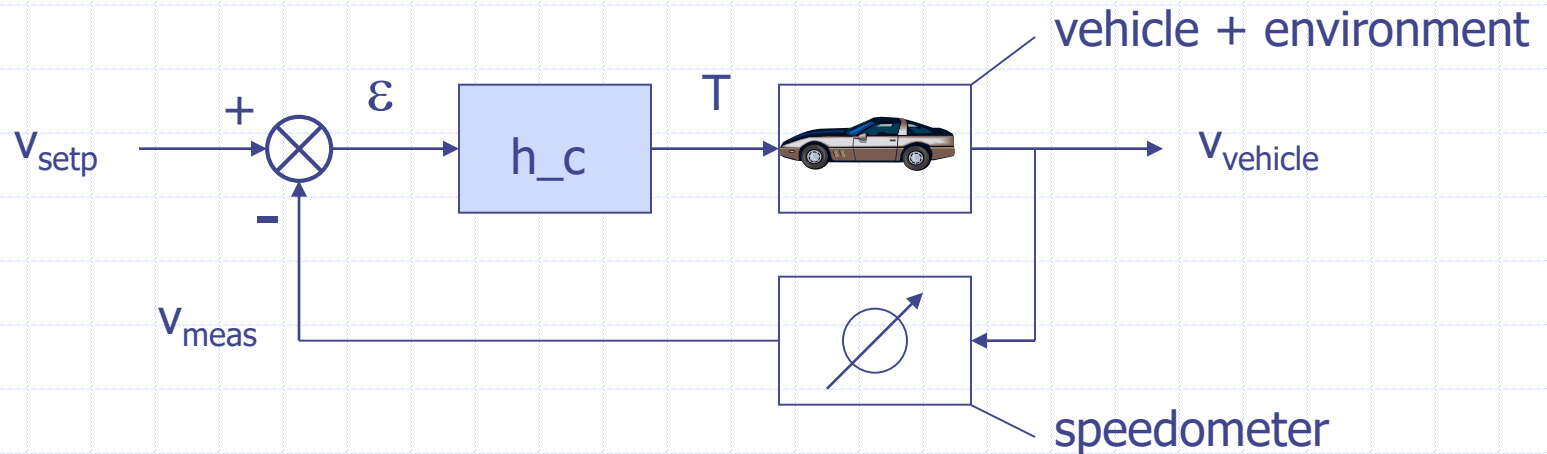
# Part I: Feedback Control

- ◆ What is Control
- ◆ The Feedback Loop
- ◆ Proportional Feedback

# Velocity Control



# Feedback Control Loop



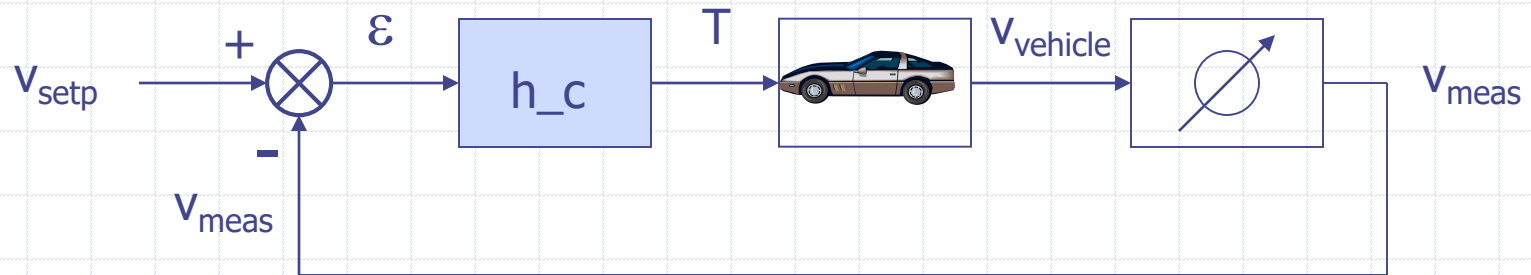
$h_c$  controller

controller function  $T = h_c(\epsilon)$ :  
adjust  $T$  such that  $\epsilon \rightarrow 0$

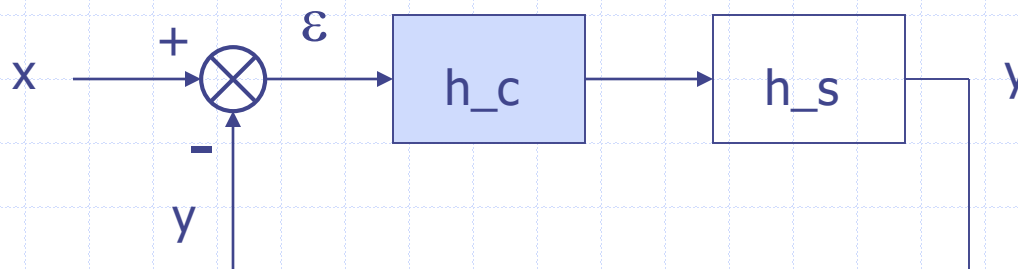
**control theory: how to determine function  $h_c$**



# Standard Loop Format



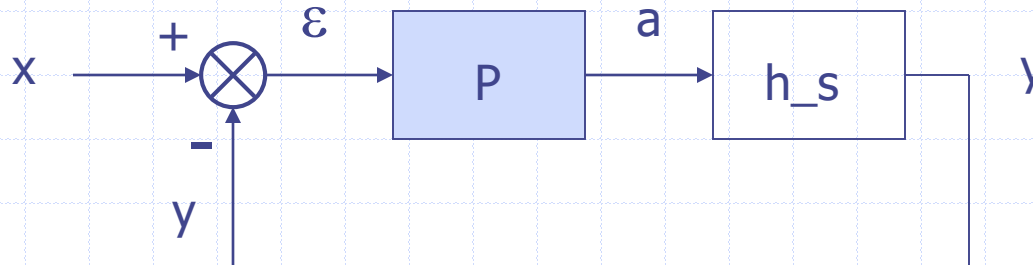
standard form: control  $h_s$  through  $h_c$  such that  $y = x$



$h_c = h_{controller}$   
 $h_s = h_{system}$

# Proportional Control

Let  $h_c(\varepsilon) = P \varepsilon$



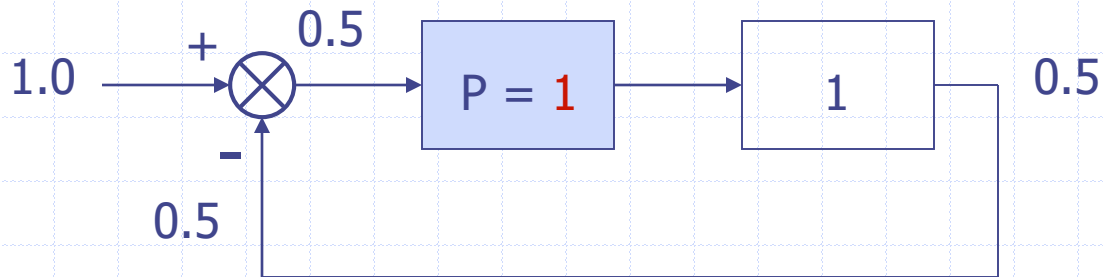
(Steady-state) Analysis:

Let  $h_s(a) = c a$  (i.e. linear system)

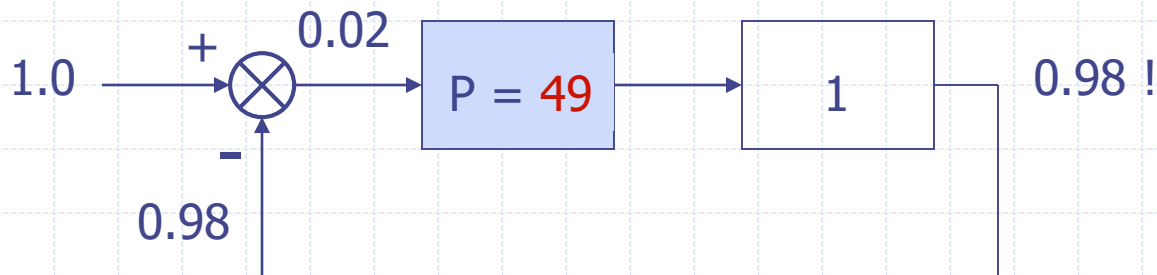
Then  $y = c P (x - y) \Rightarrow y = (c P / (c P + 1)) x$

# Effect of Loop Gain

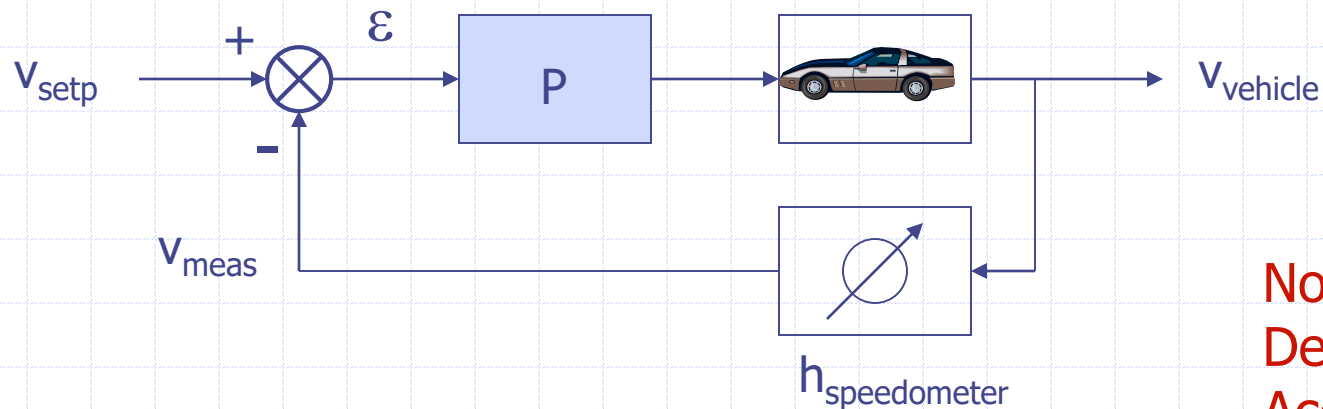
$$y = \frac{P}{P+1} x$$



Loop gain: the larger, the better ( $y \approx x$ )



# Example: Velocity Control



Note: Sensor Determines Accuracy

Analysis:

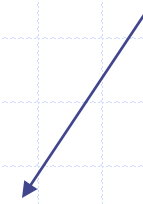
$$v_{\text{meas}} = h_{\text{speedometer}}(v_{\text{vehicle}})$$

$$\text{If } P \gg 1 \text{ then } v_{\text{meas}} \approx v_{\text{setp}}$$

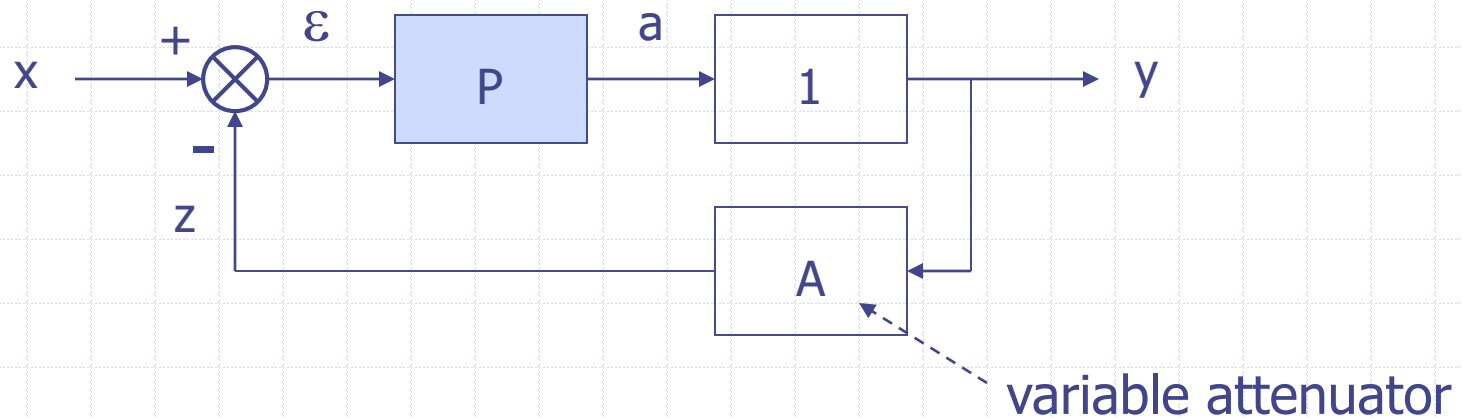
$$\text{Consequently, } v_{\text{vehicle}} \approx h_{\text{speedometer}}^{-1}(v_{\text{setp}})$$

$$\text{Ideally, } h_{\text{speedometer}}(x) = x$$

$$\text{Result: } v_{\text{vehicle}} \approx v_{\text{setp}}$$



# Example: Variable Amplifier



Analysis:

If  $PA \gg 1$  (i.e. sufficient loop gain) then  $z \approx x$

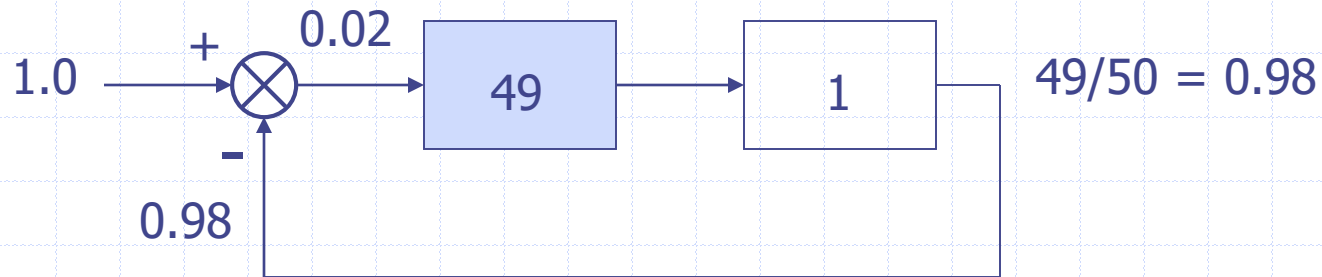
Hence  $y \approx (1/A)x$  (e.g.  $A = 0.001 \Rightarrow 1000x$  amp)

# Part II: Blessings of Feedback

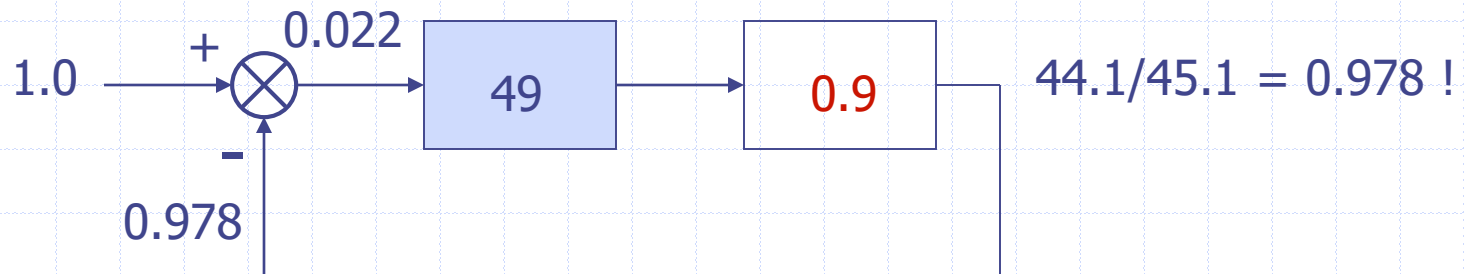
- ◆ High Loop Gain: More Robustness
- ◆ High Loop Gain: More Linearity
- ◆ High Loop Gain: More Speed

# More Robustness

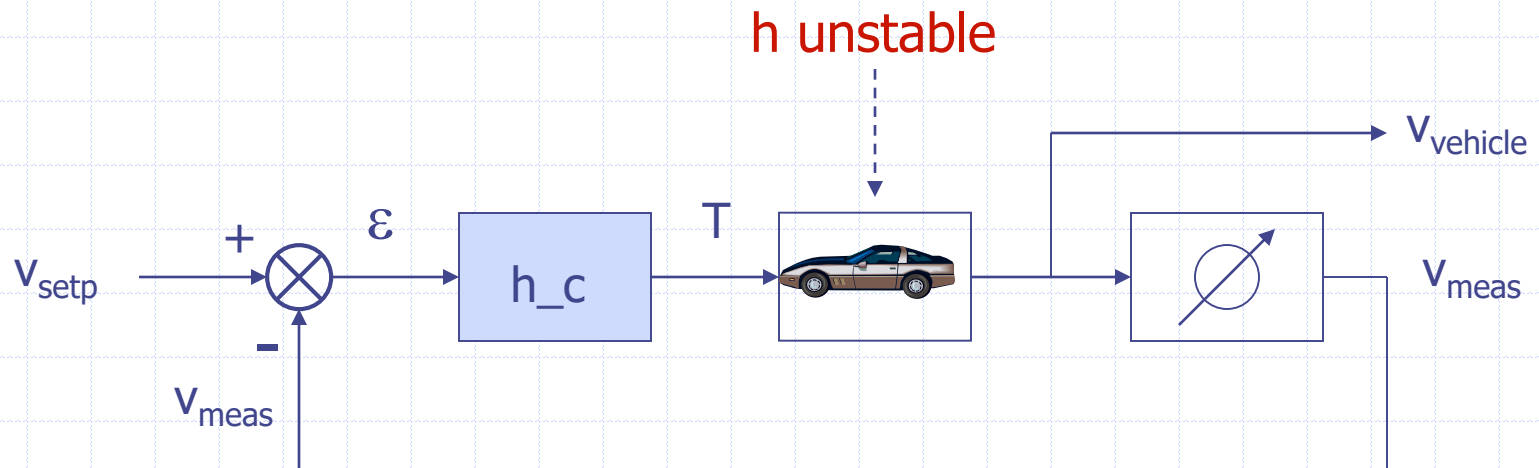
Suppose  $h_s$  varies with time



10% change in  $h_s$ : only  $10\%/50 = 0.2\%$  change in  $y$



# Example: Velocity Control

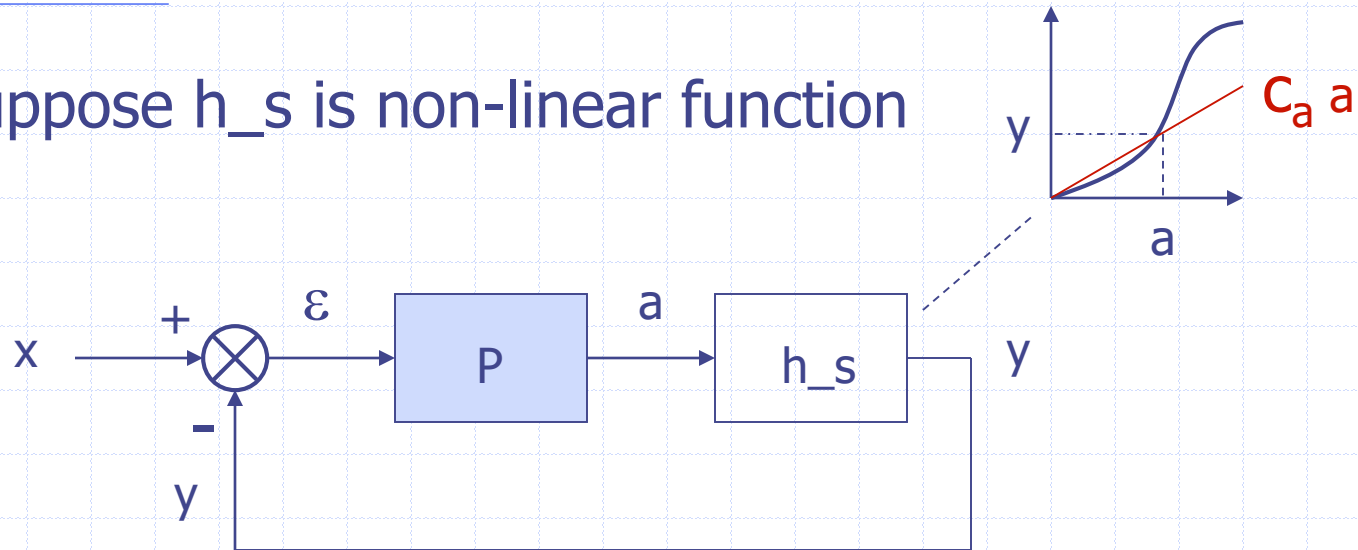


For sufficiently high loop gain:  $v_{meas}$  stable ( $\approx v_{setp}$ ),  
Hence  $v_{vehicle} \approx h_{speedometer}^{-1}(v_{setp})$ , which is stable



# More Linearity

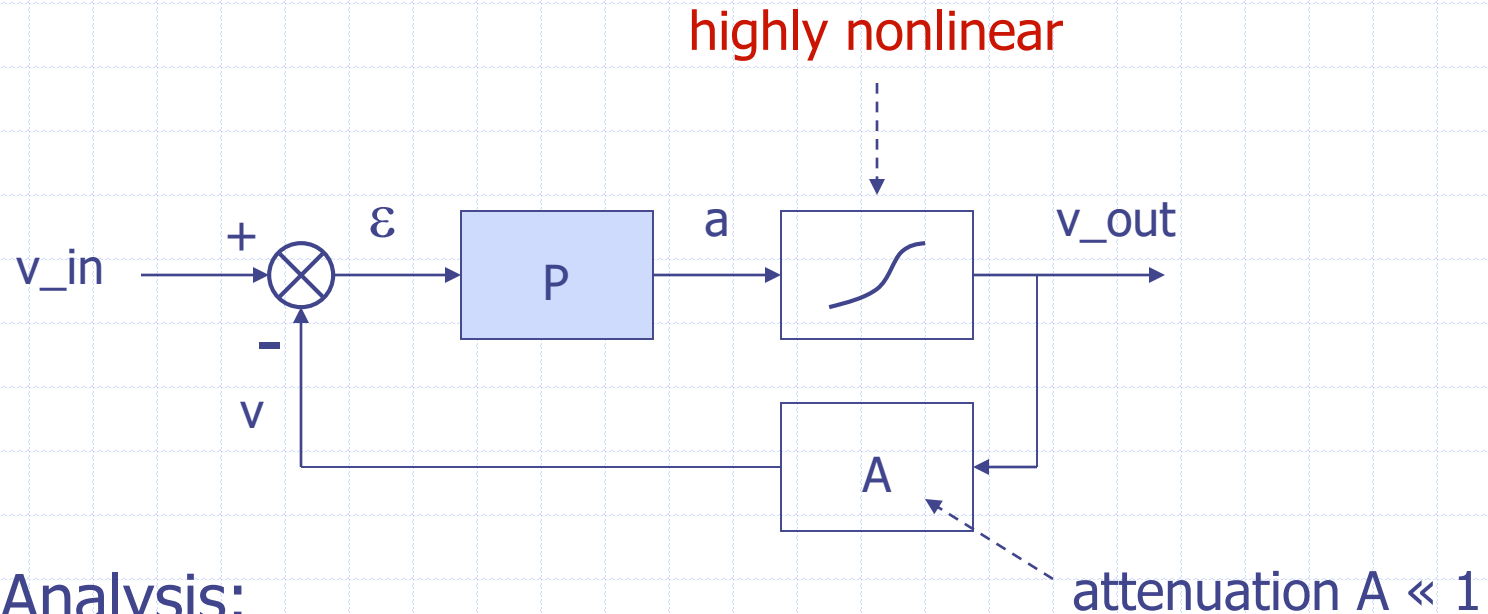
Suppose  $h_s$  is non-linear function



Analysis:

Let  $h_s(a) = c_a a \Rightarrow y = (c_a P / (c_a P + 1)) x$   
If  $c_a P \gg 1$  then  $y \approx x \Rightarrow y$  is linear with  $x$

# Example: Audio Amp

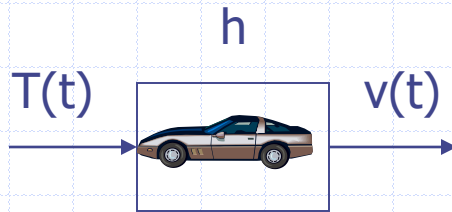


Analysis:

If  $c_a P A \gg 1$  then  $v \approx v_{in}$

Hence  $v_{out} \approx 1/A v_{in}$  (so linear gain:  $1/A$ )

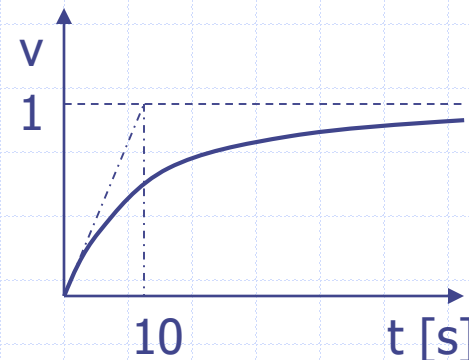
# More Speed



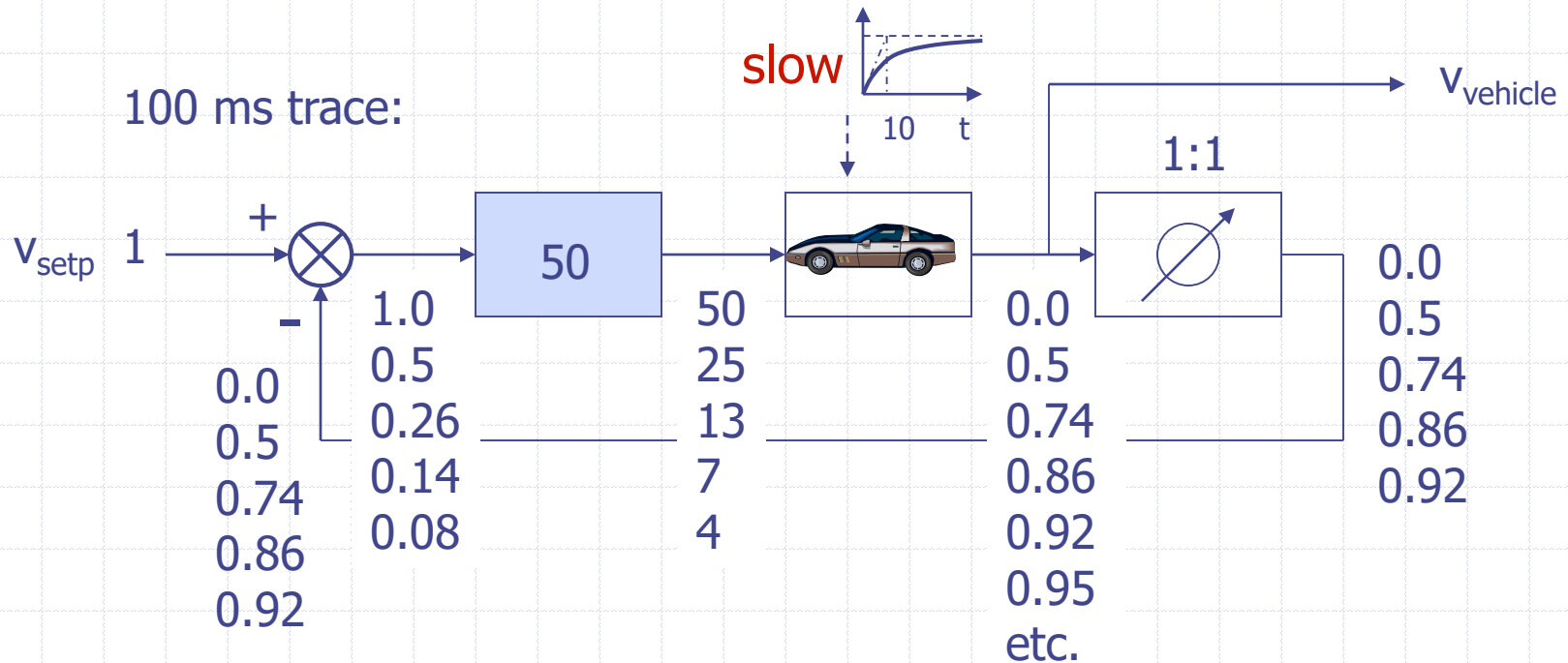
$T$  and  $v$  are typically time-varying signals (function of  $t$ ).  
transfer function ( $h$ ) is not just a proportional gain function but a first-order transfer function:

Vehicle response (slow):  
 $10(dv(t)/dt) + v(t) = T(t)$

Let  $T(t) = 1 \Rightarrow$   
 $v(t) = 1 - e^{-t/10}$



# Example: Velocity Control



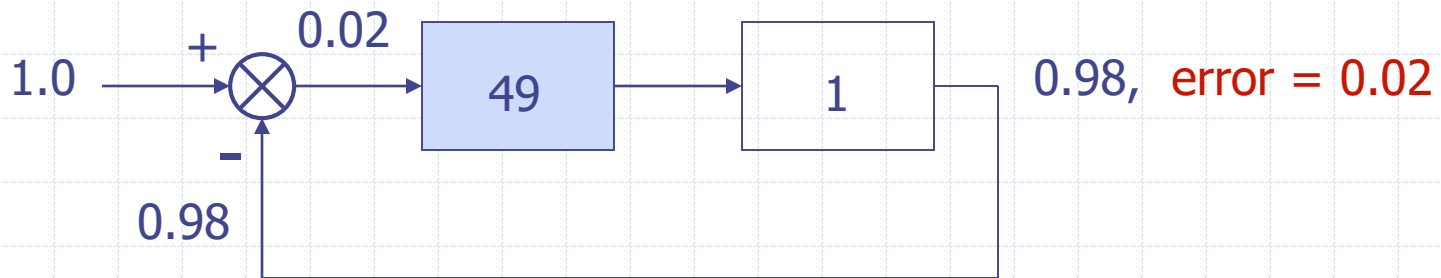
In 2 steps of 100 ms same level (0.74) as 10 s w/o feedback  
 Performance of vehicle has effectively increased  $\sim 50$  times!

# Part III: Harnessing Feedback

- ◆ Instability Problem
- ◆ Classical Control Theory

# Loop Gain Limitations

Analysis:  $y = P/(P+1) x$

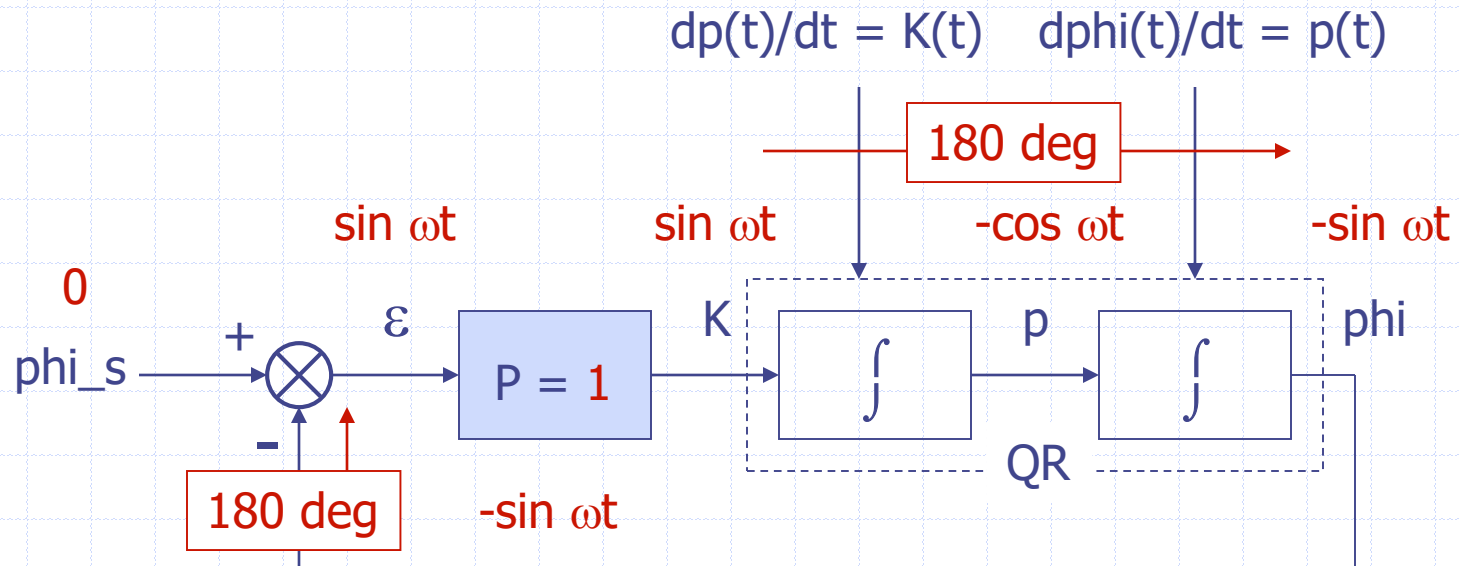


Problem:

P should be infinite for control error to become zero

In practice however, loop gain must be limited for *stability*

# Example 1: Integrator Systems

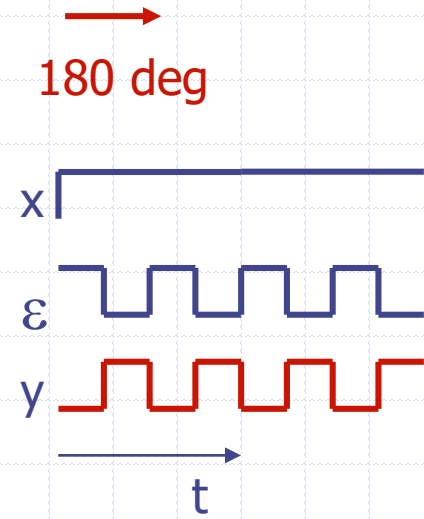
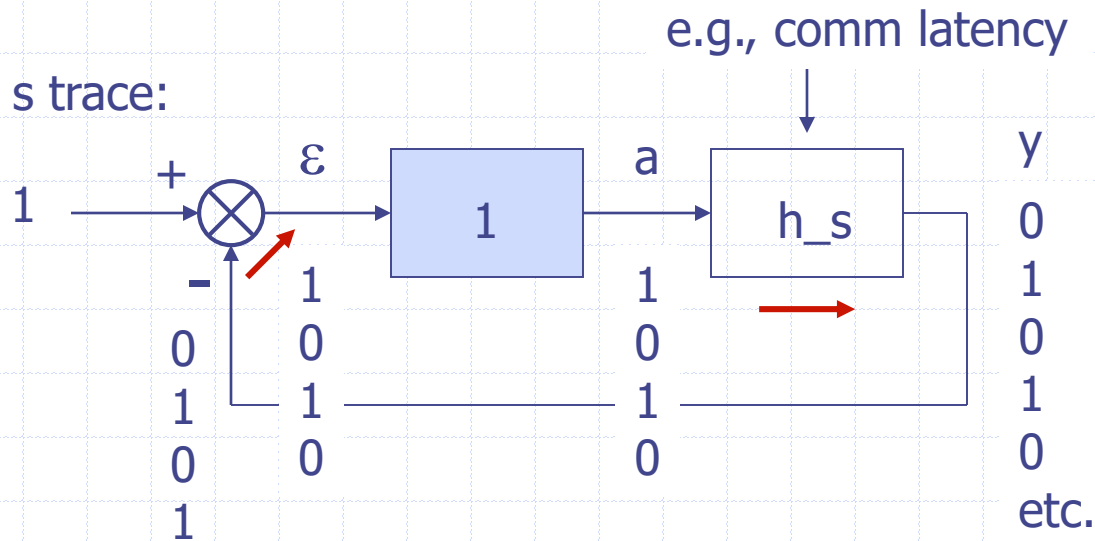


$P \geq 1$ : **instability!**

Cause: each integration adds 90 deg phase lag  
So 2 integrators use up all 180 deg budget!

# Example 2: Time Latency

0.5 s trace:



Let  $h_s: y(t) = a(t-0.5)$  (i.e., 0.5s delay)  
 Phase lag of 180 deg at 1 Hz causes **instability**



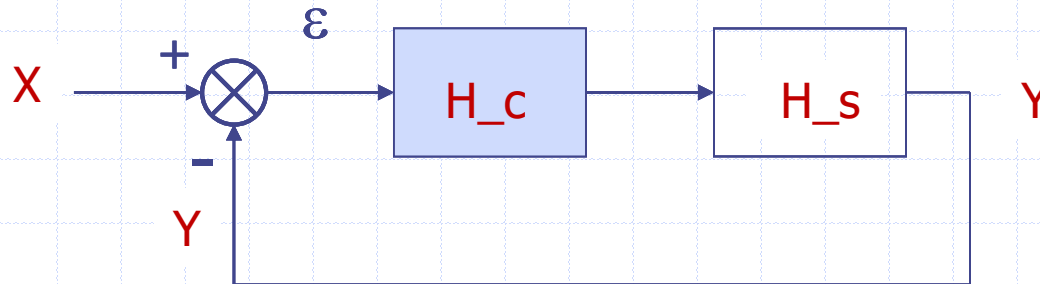
# Phase Lag: examples

$h_s$

- ◆ Integration (90 deg):
  - speed  $\rightarrow$  position, flow  $\rightarrow$  volume
- ◆ First-order system (up to 90 deg):
  - lamp, heating, car velocity, ...
- ◆ N-th order system (up to  $N \cdot 90$  deg):
  - compositions of 1st-order systems, missiles
- ◆ Delay systems (unlimited):
  - humans, computers, sample times, cables, air

Need control theory to analyze, e.g., control stability

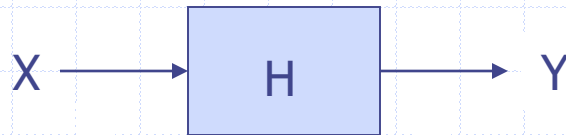
# Classical Control Theory



Describe  $x(t)$ ,  $y(t)$ ,  $h_c(t)$ ,  $h_s(t)$  in terms of their Laplace transforms  $X(s)$ ,  $Y(s)$ ,  $H_c(s)$ ,  $H_s(s)$ , respectively

$$\mathbf{L}[f(t)] = \mathbf{F}(s) = \int_0^{\infty} f(t) e^{-st} dt$$

# Classical Control Theory



For linear system  $h$  it holds  $Y(s) = H(s) \cdot X(s)$   
(i.e. composition in time domain reduces to **multiplication**  
in the Laplace domain). This allows for easy analysis.

# Laplace cheat sheet

$$\blacklozenge L[a] = a/s$$

$$\blacklozenge L[at] = a / s^2$$

$$\begin{aligned}\blacklozenge L[af + bg] &= a L[f] + b L[g] \\ &= a F(s) + b G(s)\end{aligned}$$

$$\begin{aligned}\blacklozenge L[f'] &= s L[f] - f(0) \\ &= s F(s) - f(0)\end{aligned}$$

## Laplace cheat sheet

$$\diamond L[a] = a/s$$

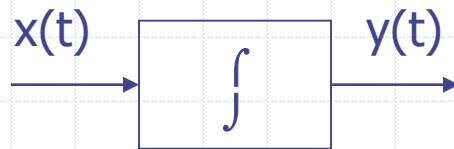
$$\diamond L[a t] = a / s^2$$

$$\diamond L[f+g] = F(s) + G(s)$$

$$\diamond L[f'] = s F(s) - f(0)$$

$$\diamond L[g(f)] = F(s) G(s)$$

# Example: Rate Control (1)



$$dy(t)/dt = x(t)$$

Laplace transform:

$$s Y(s) = X(s)$$

$$H(s) = Y(s)/X(s) = 1/s$$

$$\text{Let } x(t) = 1$$

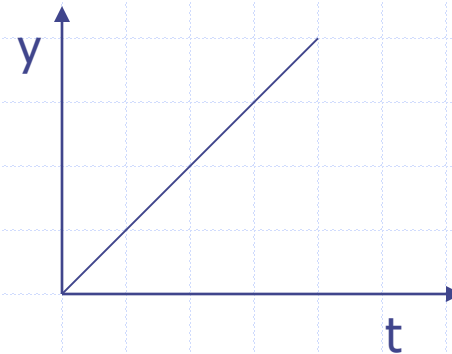
=>

$$X(s) = 1/s$$

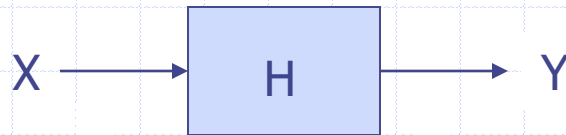
$$Y(s) = H(s) X(s) = 1/s^2$$

<=

$$y(t) = t$$



# Control System Analysis



$$Y(s) = H(s) X(s)$$

$$H(s) = \frac{Y(s)}{X(s)}$$

zeros

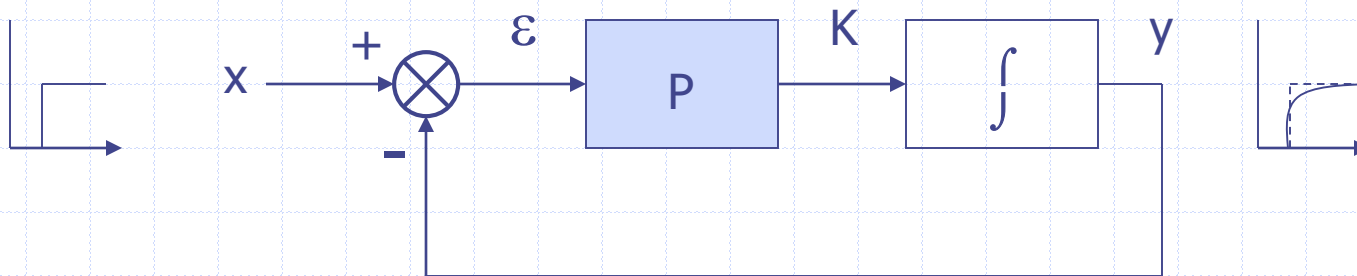
poles

Stability:

$$\text{Re}(\text{roots } H(s)) < 0$$

$\text{Im}(\text{roots } H(s))$  small

# Example: Rate Control (2)



$$Y(s) = P H(s) (X(s) - Y(s))$$

$$Y(s) = (P H(s) / (1 + P H(s))) X(s) = H_{PC}(s) X(s)$$

$$H(s) = 1/s$$

$$H_{PC}(s) = (P/s) / (1 + P/s) = P / (s + P)$$

First-order system with time constant  $1/P$

(root:  $s = -P \Rightarrow \text{Re} < 0, \text{Im} = 0$ ) so **stable**

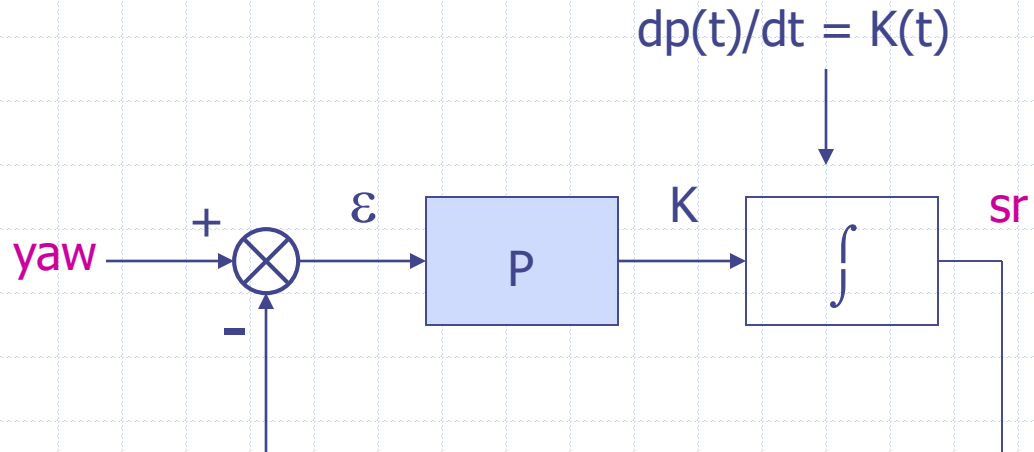
# Part IV: QR Control

- ◆ Instability Problem
- ◆ Cascaded P Control



# Rate control using P controller

P controller for roll rate:



$P < 1$ : useless control performance

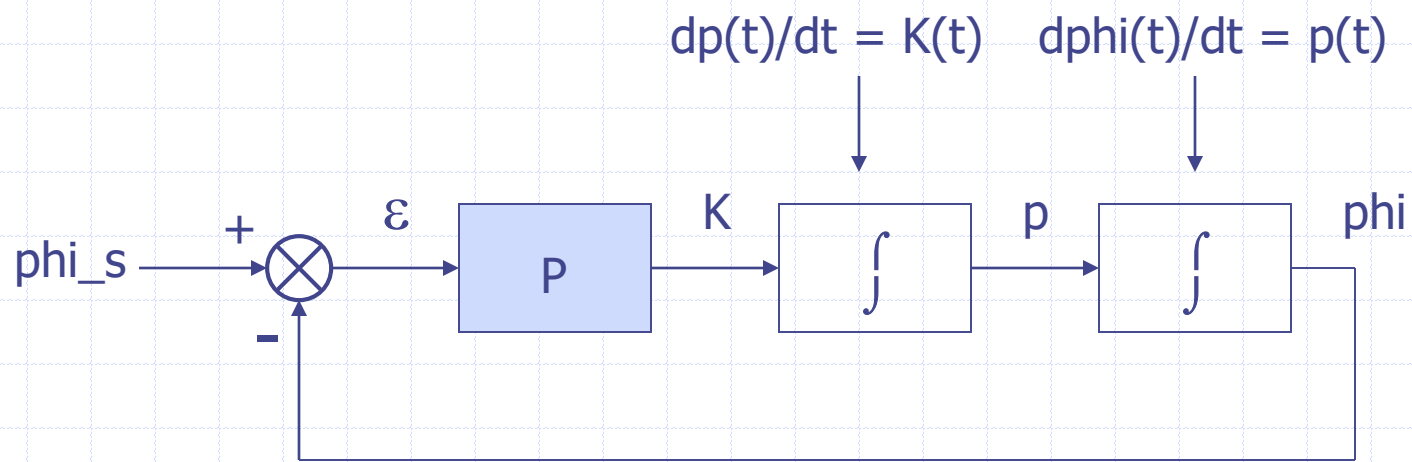
$P \geq 1$ : **stable** (for not too high P!)

## Laplace cheat sheet

- ◆  $L[a] = a/s$
- ◆  $L[at] = a / s^2$
- ◆  $L[f+g] = F(s) + G(s)$
- ◆  $L[f'] = s F(s) - f(0)$
- ◆  $L[g(f)] = F(s) G(s)$

# Angle control using P control

P controller for roll angle:



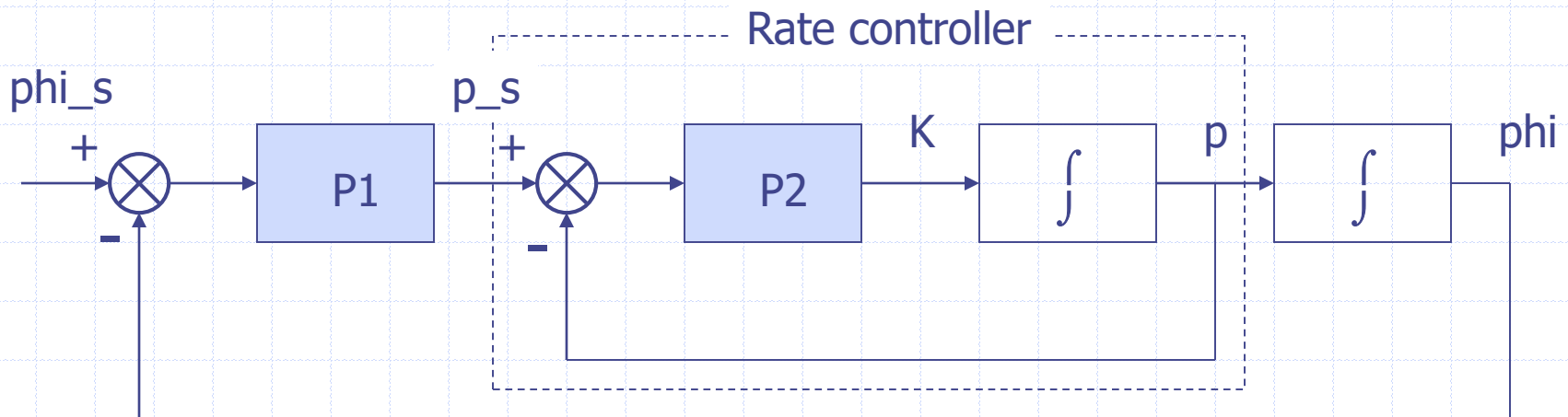
$P < 1$ : useless control performance  
 $P \geq 1$ : **instability**

## Laplace cheat sheet

- ◆  $L[a] = a/s$
- ◆  $L[at] = a / s^2$
- ◆  $L[f+g] = F(s) + G(s)$
- ◆  $L[f'] = s F(s) - f(0)$
- ◆  $L[g(f)] = F(s) G(s)$

# Cascaded P control

Embedded rate controller “neutralizes” one integrator



Cascaded P Controller: **stable** (for not too high  $P1$  and  $P2$ !  
and  $P2 \geq 4 \cdot P1$ )

# Summary

- ◆ Feedback control offers many advantages
- ◆ Is ubiquitous (cars, planes, missiles, QRs ..)
- ◆ Potential stability problems
- ◆ Need control theory
- ◆ This was merely introduction into the field
- ◆ Get a feel by applying to QR!